Heuristic-Based Algorithms for the Ordering of Gray Encoded Twiddle Factors of FFT Architectures

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Abstract

This paper addresses the exploration of different heuristic-based algorithms for a better manipulation of coefficients in twiddle factors of Fast Fourier Transform (FFT). Due to the characteristics of the FFT algorithms, which involve multiplications of input data with appropriate coefficients, the best ordering of these operations can contribute for the reduction of the switching activity, what leads to the minimization of power consumption in the FFTs. The heuristic-based algorithm named Bellmore and Nemhauser and a new proposed one named Anedma are used to get as near as possible to the optimal solution for the ordering of coefficients in FFTs with larger number of points. Since some low-power techniques for global communication in CMOS VLSI using data encoding methods are used for the decrease of power consumed for transmitting information over buses by reducing the switching activity, we have used Gray encoding technique in the coefficients. As will be shown, the appropriate ordering of coefficients, based on the guidance given by the Anedma heuristic algorithm, can contribute for the reduction of switching activity of the encoded twiddle factors.

1. Introduction

Fast Fourier Transform (FFT) is the largely implementation of the Discrete Fourier Transform (DFT), because this algorithm needs less computation due its recursive operator named butterfly. This operator performs the calculation of complex terms, which involves multiplication of input data by appropriate coefficients [1].

Coefficient ordering is used in [2] as a technique for low power, where all coefficients are ordered in a Fully-Sequential circuit so as to minimize the transitions in the multiplier input and data bus. Thus, the problem is related to finding a better ordering for each coefficient by calculating the minimum Hamming distance between the coefficients. If the number of coefficients is reduced, the cost function can be calculated for all the combinations over the coefficients, since the total number of permutations is still reasonable. However, for a higher number of coefficients this exhaustive algorithm is less attractive due to the time necessary to process the large number of combinations. In this case, heuristic algorithms should be used to get as near as possible to the optimal solution. In this paper, we explore the use of different heuristic-based algorithms in order to search the best ordering of the coefficients for the power reduction of sequential FFT architectures. Two heuristic algorithms are used for this purpose, where one of them is named Bellmore and Nemhauser [3] and the other one is a new proposed heuristic named Anedma [4], which were implemented in order to get as near as possible to the optimal solution for the ordering of larger FFT instances. The main results show that based on the guidance given by the Anedma heuristic algorithm, the switching activity of the twiddle factors can be reduced significantly.

Since the data encoding method has been used for the decrease of power consumed over buses, by reducing the switching activity, we have used Gray encoding technique in order to verify the impact on reducing the switching activity of the coefficients. The results show that depending on the heuristic-based algorithm used, the number of transitions in the encoded twiddle factors can be reduced considerably.

2. FFT Structure

The most popular among the algorithms for the FFT calculation is named common factor FFT [5]. In this work, the main focus is the radix-2 common factor algorithm with decimation in frequency. In this algorithm, the frequency samples are decimated during each stage of the FFT. The operations are realized at each pair of input signals. The basic structure of this algorithm is shown in Fig. 1 for a 16-point example, where the computational structure of the FFT is divided into stages, groups and butterflies. The amount of stages of a radix-2 common factor FFT is given by log2N, where N represents the number of points of the FFT.

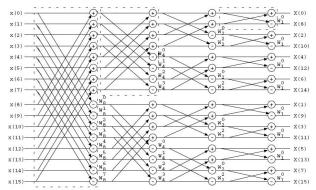


Fig. 1 - Data flow for 16-point radix-2 common factor FFT with decimation in frequency

In the structure presented in Fig. 1, 32 real and 32 imaginary terms are performed in the butterfly (4 stages with 8 butterflies). The butterfly plays a central role in the FFT computation. For the common factor FFT algorithm with decimation in frequency, the butterfly allows the calculation of complex terms according to eq. (1) and (2).

$$C_{complex} = A_{complex} + B_{complex} \tag{1}$$

$$C_{complex} = A_{complex} + B_{complex}$$

$$D_{complex} = (A_{complex} - B_{complex}) W_{complex}$$
(1)

As can be observed in the equations above, one complex addition, one complex subtraction and one complex multiplication are involved in the butterfly block. In the complex multiplication, the input samples are multiplied by fixed coefficients named twiddle factors (W) These coefficients represent values multiple of

 $W_N = e^{-j\frac{2\pi K}{N}} = \cos\frac{2\pi k}{N} - j\sin\frac{2\pi k}{N}$, where k is the level of the butterfly at factor, and they are represented by: each stage of the FFT.

2.1 Ordering of the Twiddle Factors

Tab.1 shows an example of the calculation of the first column (stage) of the FFT, based on the structure previously presented in Fig. 1. This example considers a fully sequential structure with both normal (sequential with no ordering) and after using ordering of the coefficients. In this work we use Gray encoding technique for both i) verify the impact on reducing the switching activity in the original FIR filter coefficients, and ii) verify the best ordering of the encoded coefficients when using the heuristic-based algorithms.

Tab. 1 - Ordering of coefficients in the FFT algorithm Clock cycles Normal operation Operation after ordering the coefficients xo + x8Xo + x8 $(x0-x8) \times W_8^0$ $(x0-x8) \times W_8^0$ 2 x1 + x9x6 + x14 $(x1 - x9) x W_8$ $(x6 - x14) \times W_8^6$ 3 x2 + x10 $(x2 - x10) \times W_8^2$ $(x2 - x10) \times W_8$ 4 x3 + x11x5 + x13 $(x3 - x11) \times W_8^3$ $(x5 - x13) \times W_8^5$ x4 + x12x3 + x11 $(x3 - x11) \times W_8^3$ $(x4 - x12) \times W_8^4$ 6 x5 + x13 $(x5 - x13) \times W_8$ $(x1 - x9) \times W_8$ x6 + x14x4 + x12 $(x6 - x14) \times W_8^6$ $(x4 - x12) \times W_8^4$ x7 + x15x7 + x15 $(x7 - x15) \times W_8^7$ $(x7 - x15) \times W_8^7$

3. Heuristic-based algorithms

This section summarizes the main aspects of the heuristic-based algorithms used in this work.

3.1. Bellmore and Nemhauser Algorithm

Tab.2 presents a random example with the symmetric cost to add each element to the sequence of decisions. Considering the same element values from Tab.2, we would have the following solution ([E2] [E6] [E1] [E3] [E4] [E5]), which would be formed by the steps shown in Fig. 2 (starting from the initial element E1).

Element	E1	E2	E3	E4	E5	E6
E1	0	2	1	4	9	1
E2	2	0	5	9	7	2
E3	1	5	0	3	8	6
E4	4	9	3	0	2	5
E5	9	7	8	2	0	2
E6	1	2	6	5	2	0
Step 2: Insert E6	because it is an	nong the elem	ents not yet a		g the best neigh	
Step 2: Insert E6 the list, E6 appears ighbor of E3 with	because it is an	nong the elemighbor of E1,	ents not yet a with a cost o	dded. Choosing f 1. At the oth	g the best neigh er end of the li	st, E4 is tl
Step 2: Insert E6 the list, E6 appears ghbor of E3 with	because it is an s as the best ne a cost of 3. Th	nong the elemighbor of E1, en, E6 is add	ents not yet a with a cost o ed to the extr	dded. Choosing f 1. At the oth reme left of the	g the best neighter end of the list route. Current	st, E4 is th t list: ([E6
Step 2: Insert E6 the list, E6 appearing the E3 with [3]).	because it is an a sas the best ne a cost of 3. The at the left end of	nong the elemighbor of E1, en, E6 is add	ents not yet a with a cost o ed to the extrem the element	dded. Choosing f 1. At the oth reme left of the swhich are no	g the best neighter end of the list route. Current tadded yet, the	st, E4 is t t list: ([E6 best neig
Step 2: Insert E6 the list, E6 appears ghbor of E3 with B]). Step 3: Insert E2 is E2 with a cost	because it is and as as the best neal cost of 3. The at the left end of 2, and the electrons are the second as the second at the left end of 2, and the electrons are the second as the second as the second are the second as the second are the seco	nong the elemighbor of E1, en, E6 is add of the list. From the ement with lo	ents not yet a with a cost o ed to the extrem the element wer cost from	dded. Choosing f 1. At the oth reme left of the s which are no a E3 is E4 with	g the best neighter end of the list route. Current added yet, the cost 3. Current	st, E4 is t t list: ([E6 best neig t list: ([E2

Step 5: As the only remaining element is the E5, we just must see which end their inclusion has lower cost. If E5 is inserted at the end of E2, then the cost is equal to 7. On the other hand, if inserted as a neighbor of E4 then the cost is 2. Final list: ([E2] [E6] [E1] [E3] [E4] [E5]).

The total cost of the elements of the list is given by:

TotalCost = cost(E2, E6) + cost(E6, E1) + cost(E1, E3) + cost(E3, E4) + cost(E4, E5) + cost(E5, E2) = 16.

Fig. 2 – Bellmore and Nemhauser steps

3.2. Anedma Algorithm

By using this heuristic, the purpose is that a coefficient which has P candidates with the same Hamming distance to be chosen as a neighbor, make the choice of the lowest important coefficient for the others neighbors, thus making it possible to achieve better results. In this heuristic, P threads are started and each one produces an ordering resulting from the application of the algorithm using a random point as initial element. In Tab.3, a hypothetic example with 10 elements randomly generated representing the coefficients and the Hamming distance between them. The total Hamming distance with lexicographical ordering (E1,... ,E10) is equal to 27. The list of candidates of each element, considering E1 as the initial element is shown in Fig. 3.

Elements	E1	E2	E3	E4	E5	E6	E7	E8	E9	E10
E1	0	2	1	4	9	1	1	3	8	1
E2	2	0	5	9	7	2	3	2	2	5
E3	1	5	0	3	8	6	8	9	2	3
E4	4	9	3	0	2	5	1	6	7	2
E5	9	7	8	2	0	2	5	3	2	8
E6	1	2	6	5	2	0	3	2	4	3
E7	1	3	8	1	5	3	0	7	6	2
E8	3	2	9	6	3	2	7	0	1	6
E9	8	2	2	7	2	4	6	1	0	2

Tab. 3 – A hypothetic example with 10 coefficients

E1C = ([E3], [E6], [E7], [E10]);	E1C = ([E3], [E6], [E10]);		E1C = ([E3], [E10]);	
E2C = ([E6], [E8], [E9]);	E2C = ([E6]);		E2C = ([E6]);	
E3C = ([E9]);	E3C = ([E9]);		E3C = ([E9]);	
E4C = ([E7]);	E4C = ([E7]);		E4C = ([E7]);	
E5C = ([E4], [E6], [E9]);	E5C = ([E6]);		E5C = ();	
E6C = ([E2], [E5], [E8]);	E6C = ([E2], [E5]);		E6C = ([E2], [E5]);	
E7C = ([E4]);	E7C = ();	E8C = ();	E7C = ();	E8C = ();
E8C = $([E9]);$	E9C = ([E8]);	E10C = ([E4]);	E9C = ([E8]);	E10C = (E4).
E9C = ($[E8]$);	Finally, the list EBloo	ı is given by::	Ebloq = ([E1], [E9], [E9])	E7], [E4], [E8],
E10C = ([E4]).	Ebloq=([E1],[E9],[E	7],[E4],[E8])	[E6])	
(a)	(b)		(c)	

Fig. 3. Steps of the Anedma algorithm

E10	1	5	3	1	8	3	2	6	2	0

In addition, E1 is already entered in a locked elements list named EBloq that now is EBloq = ([E1]). This list prevents that these elements will be selected as candidates from other factors. E1C represents the list of elements candidate of (coefficient) E1, E2C represents the candidate list of element E2, and successively. Fig. 3(a) shows this procedure. The partial Ebloq list is shown in Fig. 3(b). The final Ebloq list, with the total Hamming of this ordering reduced to 20 is shown in Fig. 3(c).

4. Results

In this section, we present the results obtained with the two heuristic-based algorithms and with the encoding technique described in the previous sections. We used 16 bit-width FFT architectures with 128, 256, 512 and 1024 points. As the twiddle factors are composed by complex terms, the coefficients were divided into real and imaginary parts.

Tab.4 shows the obtained results, in terms of number of transitions, after applying Gray encoding in the original and ordered coefficients (after applying Bellmore and Nemhauser and Anedma algorithms). As should be seen, both algorithms achieved significant reduction in terms of Hamming distance, mainly on the set of real coefficients with the Anedma algorithm. It occurs because this algorithm uses a list of candidates that provides (when more than one coefficient with the same Hamming of the best candidate is found), the possibility of not making a random choice between them, but keep them in a candidate list. In case of any of these elements is the unique best candidate to one of the following coefficients analyzed, it can be removed from this list to be added as best candidate of another element. Thus, each coefficient with more than one candidate always will select the candidate that is the one which has the less importance to others coefficients, thus providing further optimization of the obtained results.

1 ab. 4 – Applying the ordering on the Gray encoded coefficients											
	Number of transitions between the coefficients (Hamming distance)										
			Ordered Real Coefficients				Ordered Imaginary Coefficients				
	Original Coefficients		Bellmore and		Anedma		Bellmore and		Anedma		
Twiddles factors size	Real	Imaginary	Nemhauser				Nemhauser				
	Ham.	Ham.	Ham	Red. (%)	Ham	Red. (%)	Ham	Red. (%)	Ham	Red. (%)	
128	508	232	206	59,4	200	60,6	142	38,7	138	40,5	
256	926	436	376	59,4	372	59,8	258	40,8	250	42,7	
512	1702	816	700	58,9	694	59,2	446	45,3	440	46,1	
1024	3138	1542	1248	60,2	1246	60,3	748	51,4	740	52,0	

 Γ ab. 4 – Applying the ordering on the Gray encoded coefficients

5. Conclusions

In this work two heuristic-based algorithms named Anedma and Bellmore and Nemhauser were applied to the ordering of Gray encoded FFT twiddle factors. The results shows that the algorithms can find a good cost function for the ordering of the coefficients and the Hamming distance between consecutive coefficients could be reduced significantly. As future work we intend to implement the FFT architectures with original and encoded ordered coefficients in order to observe the impact on power reduction of the FFT with the reduction of the number of transitions in the coefficients.

6. References

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